Basics of Analytic Signals

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May 17, 2023

Motivation

- Many natural and man-made signals exhibit *time-varying frequencies* (e.g., chirps, FM radio waves).
- Characterization and analysis of such a signal, u(t), based on instantaneous amplitude a(t), instantaneous phase $\phi(t)$, and instantaneous frequency $\omega(t) := \phi'(t)$, are very important:

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- It is convenient to use a complexified version of the signal whose real part is a given real-valued signal u(t).
- Given u(t), however, there are infinitely many ways to define the instantaneous amplitude and phase (IAP) pairs so that

$$u(t) = a(t)\cos\phi(t).$$

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$$f(t) = u(t) + iv(t)$$

where v(t) is an arbitrary real-valued signal; yet this yields the IAP representation of u(t) via

$$a(t) = \sqrt{u^2(t) + v^2(t)}, \quad \phi(t) = \arctan \frac{v(t)}{u(t)}.$$

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 - u(t) must be derived from u(t).
 - 2 Amplitude continuity: a small change in $u \Rightarrow$ a small change in a(t)
 - 3 Phase independence of scale: if cu(t), $c \in \mathbb{R}$ arbitrary scalar, then the phase does not change from that of u(t) and its amplitude becomes c times that of u(t).
 - Harmonic correspondence: if $u(t) = a_0 \cos(\omega_0 t + \phi_0)$, then $a(t) \equiv a_0$, $\phi(t) \equiv \omega_0 t + \phi_0$.

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- For simplicity, we assume that our signals are 2π -periodic in $\theta \in [-\pi,\pi)$.
- Hence, we work on the unit circle and unit disk \mathbb{D} in $\mathbb{C} = \mathbb{R}^2$
- Note that the signals over $\mathbb{R} = (-\infty, \infty)$ can be treated similarly by considering the real axis and the upper half plane of \mathbb{C} .
- The analytic signal of a given signal $u(\theta) \in \mathbb{R}$ is often and simply obtained via the *Hilbert transform*:

$$f(\theta) = u(\theta) + i\mathcal{H}u(\theta), \quad \mathcal{H}u(\theta) := \frac{1}{2\pi} \operatorname{pv} \int_{-\pi}^{\pi} u(\tau) \cot \frac{\theta - \tau}{2} d\tau.$$

Note that

$$u(\theta) = \frac{a_0}{2} + \sum_{k \ge 1} (a_k \cos k\theta + b_k \sin k\theta) \Rightarrow \mathcal{H}u(\theta) = \sum_{k \ge 1} (a_k \sin k\theta - b_k \cos k\theta).$$

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We can gain a deeper insight by viewing this as the boundary value of an analytic function F(z) where

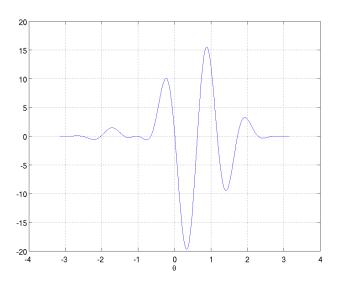
$$F(z) := U(z) + i\widetilde{U}(z), \quad z \in \mathbb{D},$$

where

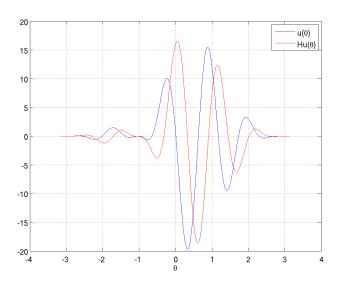
$$\begin{split} U(z) &= U\left(r\mathrm{e}^{\mathrm{i}\theta}\right) = P_r * u(\theta) = \frac{1}{2\pi} \int_{-\pi}^{\pi} \frac{1 - r^2}{1 - 2r\cos(\theta - \tau) + r^2} u(\tau) \,\mathrm{d}\tau, \\ \widetilde{U}(z) &= \widetilde{U}\left(r\mathrm{e}^{\mathrm{i}\theta}\right) = Q_r * u(\theta) = \frac{1}{2\pi} \int_{-\pi}^{\pi} \frac{2r\sin(\theta - \tau)}{1 - 2r\cos(\theta - \tau) + r^2} u(\tau) \,\mathrm{d}\tau. \end{split}$$

In other words, the original signal $u(\theta) = U(e^{i\theta})$ is the boundary value of the harmonic function U on $\partial \mathbb{D}$, which is constructed by the Poisson integral. \widetilde{U} and $Q_r(\theta)$ are referred to as the conjugate harmonic function and the conjugate Poisson kernel, respectively.

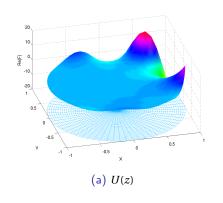
Analytic Signal . . . An Example: $u(\theta)$

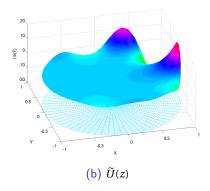


Analytic Signal . . . An Example: $u(\theta)$ and $\mathcal{H}u(\theta)$



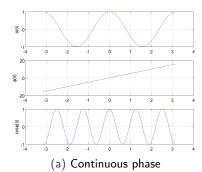
Analytic Signal . . . An Example: U(z) and $\widetilde{U}(z)$





Even if we use the analytic signal, its IAP representation is not unique as shown by Cohen, Loughlin, and Vakman (1999):

- $f(\theta) = a(\theta)e^{i\phi(\theta)}$, where $a(\theta) = u(\theta)\cos\phi(\theta) + v(\theta)\sin\phi(\theta)$ may be negative though $\phi(\theta)$ is continuous;
- $f(\theta) = |a(\theta)|e^{i(\phi(\theta) + \pi\alpha(\theta))}$, where $\alpha(\theta)$ is an appropriate phase function which may be discontinuous.



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