

Lecture 8: Fast Fourier Transform (FFT)

Note Title

Introductory Remarks:

* Various programs exist, e.g., MATLAB, R, Mathematica, ..., as well as the public domain source codes, e.g., FFTPACK @ netlib, ...

Perhaps, the most popular one is:

FFTW available from <http://www.fftw.org>

* It's known by Gauss!
See the article by M. Heideman et al. (1985).

* $W_N^* f$ (forward) or $W_N F$ (inverse) cost $O(N^2)$ if you use the conventional matrix-vector multiplications.
⇒ Too expensive for large N .

* W_N has a very **special structure** and FFT algorithms fully utilize that to achieve **$O(N \log_2 N)$** operations.

In this lecture, we use the following notation / convention for convenience:

$$F[k] = \mathcal{D}_N\{f\}[k] = \sum_{l=0}^{N-1} f[l] W_N^{-kl}$$

In other words, we use the notation $f[l]$ for f_l , $F[k]$ for F_k , $l, k = 0, 1, \dots, N-1$, there's no normalizing const. $\frac{1}{\sqrt{N}}$.

(We can always normalize by $\frac{1}{\sqrt{N}}$ later.)

Note first:

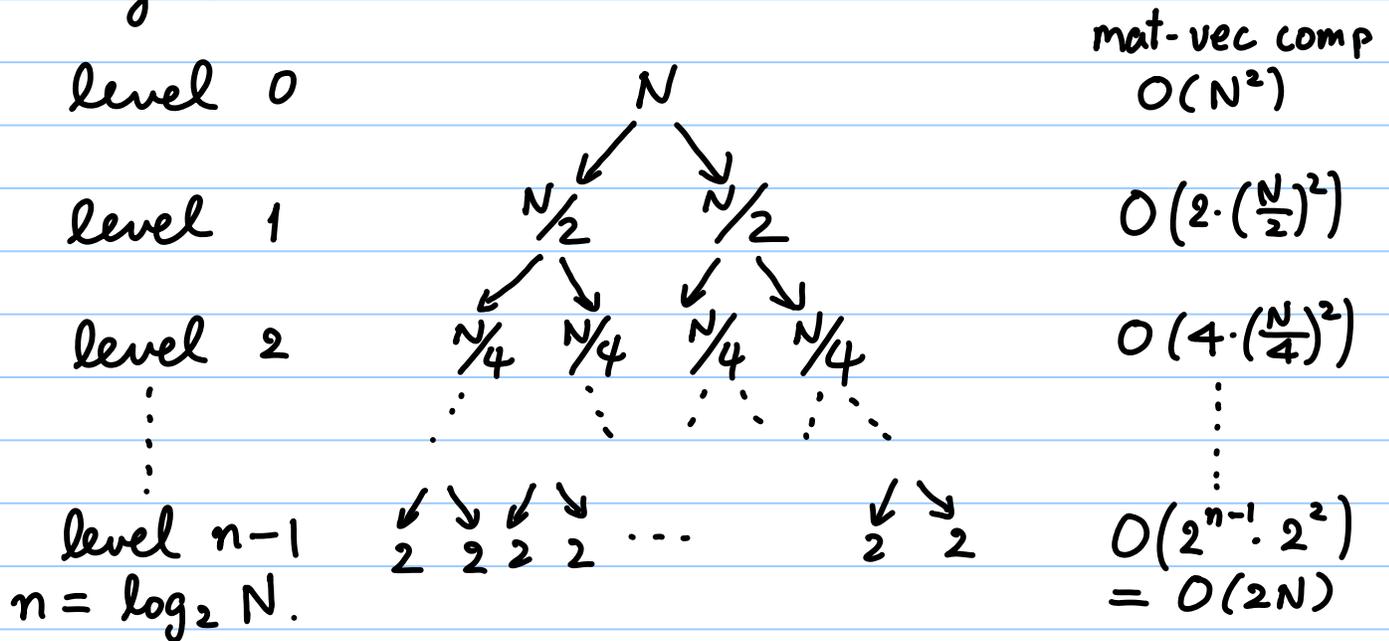
$$\omega_N^{(k+mN) \cdot (l+nN)} = \omega_N^{kl} \text{ by periodicity}$$
$$\omega_N^{k \cdot l} \in \{ \underbrace{\omega_N^0}_{=1}, \omega_N^1, \dots, \underbrace{\omega_N^{N/2}}_{=-1}, \dots, \omega_N^{N-1} \}$$

⇒ For each fixed N , we can store these N complex numbers as a **table** once & for all!

★ The Basic Idea of FFT (Cooley-Tukey, 1965)

It's **hierarchical** in nature (V. Rokhlin's comment: "it's created by God.")

Say $N = 2^n$.



But in reality, we cannot decrease the complexity by $1/2$ at each level.

≡ many variants of FFT. We will only describe the following.

★ Decimation in Time (DIT) Algorithm

Suppose $N = 2^n$ as usual.

The heart of the matter:

Split an input seq. into two sub-seq.'s of even & odd indices of the original!

- Define two subseq.'s of $\{f[l]\}_{l=0}^{N-1}$:

$$\begin{cases} f_0[l] := f[2l] \\ f_1[l] := f[2l+1] \end{cases} \quad l = 0, 1, \dots, \frac{N}{2}-1.$$

- Then we have

$$F[k] = \sum_{l=0}^{N-1} f[l] \omega_N^{-kl} = \sum_{l=0}^{\frac{N}{2}-1} f[2l] \omega_N^{-k \cdot 2l} + \sum_{l=0}^{\frac{N}{2}-1} f[2l+1] \omega_N^{-k(2l+1)}$$

$$= e^{-\frac{2\pi i k \cdot 2l}{N}} \sum_{l=0}^{\frac{N}{2}-1} f_0[l] \omega_{N/2}^{-kl} + \omega_N^{-k} \sum_{l=0}^{\frac{N}{2}-1} f_1[l] \omega_{N/2}^{-kl}, \quad k=0, 1, \dots, N-1$$

$$= e^{-\frac{2\pi i k l}{N/2}}$$

$$= \omega_{N/2}^{-kl}$$

$$= \mathcal{D}_{N/2}\{f_0\}[k]$$

$N/2$ pt DFT

$$=: F_0[k]$$

$$= \mathcal{D}_{N/2}\{f_1\}[k]$$

$N/2$ pt DFT

$$=: F_1[k]$$

Now, note that F_0 & F_1 are both $N/2$ -periodic.

Hence we only need to record
 $F_0[0] \sim F_0[\frac{N}{2}-1]$ and $F_1[0] \sim F_1[\frac{N}{2}-1]$
 i.e., N complex numbers!

In fact,

This is called \Rightarrow the butterfly relation

$$F[k] = \begin{cases} F_0[k] + \omega_N^{-k} F_1[k], & k=0, 1, \dots, \frac{N}{2}-1. \\ F_0[k-\frac{N}{2}] - \omega_N^{-(k-\frac{N}{2})} F_1[k-\frac{N}{2}], & k=\frac{N}{2}, \dots, N-1. \end{cases}$$

$$\begin{aligned} \textcircled{\smile} & F_0[k-\frac{N}{2}] + \omega_N^{-k} F_1[k-\frac{N}{2}], \quad k=\frac{N}{2}, \dots, N-1 \\ & = F_0[k-\frac{N}{2}] + \omega_N^{-(k-\frac{N}{2})} \omega_N^{-\frac{N}{2}} F_1[k-\frac{N}{2}] \\ & = F_0[k-\frac{N}{2}] - \omega_N^{-(k-\frac{N}{2})} F_1[k-\frac{N}{2}] \quad \text{since } \omega_N^{-\frac{N}{2}} = -1 \end{aligned}$$

This butterfly op. can also be written as

$$\begin{cases} F[k] = F_0[k] + \omega_N^{-k} F_1[k] & k=0, 1, \dots, \frac{N}{2}-1 \\ F[k+\frac{N}{2}] = F_0[k] - \omega_N^{-k} F_1[k] & \text{same!} \end{cases}$$

\hookrightarrow requires 1 complex mult. + 2 complex additions.

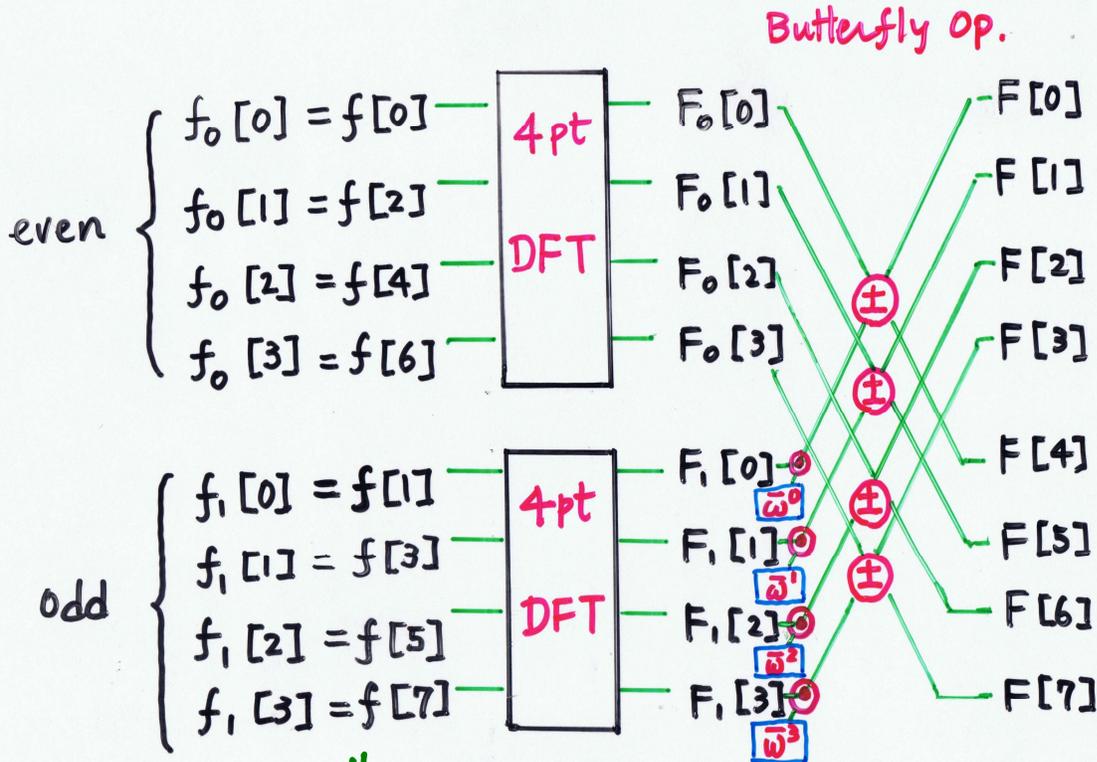
- This split procedure is repeated recursively.

- At the bottom level $n-1$:

$$\omega_2 = e^{\frac{2\pi i}{2}} = -1. \quad \begin{cases} F[0] = f[0] + \omega_2^0 f[1] = f[0] + f[1] & \text{sum} \\ F[1] = f[0] + \omega_2^1 f[1] = f[0] - f[1] & \text{difference} \end{cases}$$

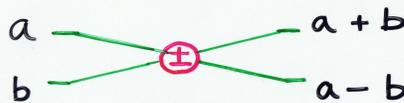
Let's look at an 8-pt FFT as an example!

1 level FFT (N=8)

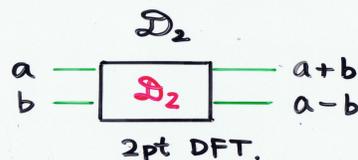
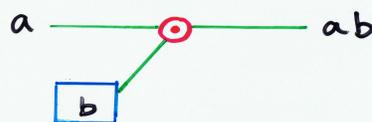


↓
 This order of the original seq. becomes the so-called **bit-reversal order** if we go down to the bottom level $n-1$.

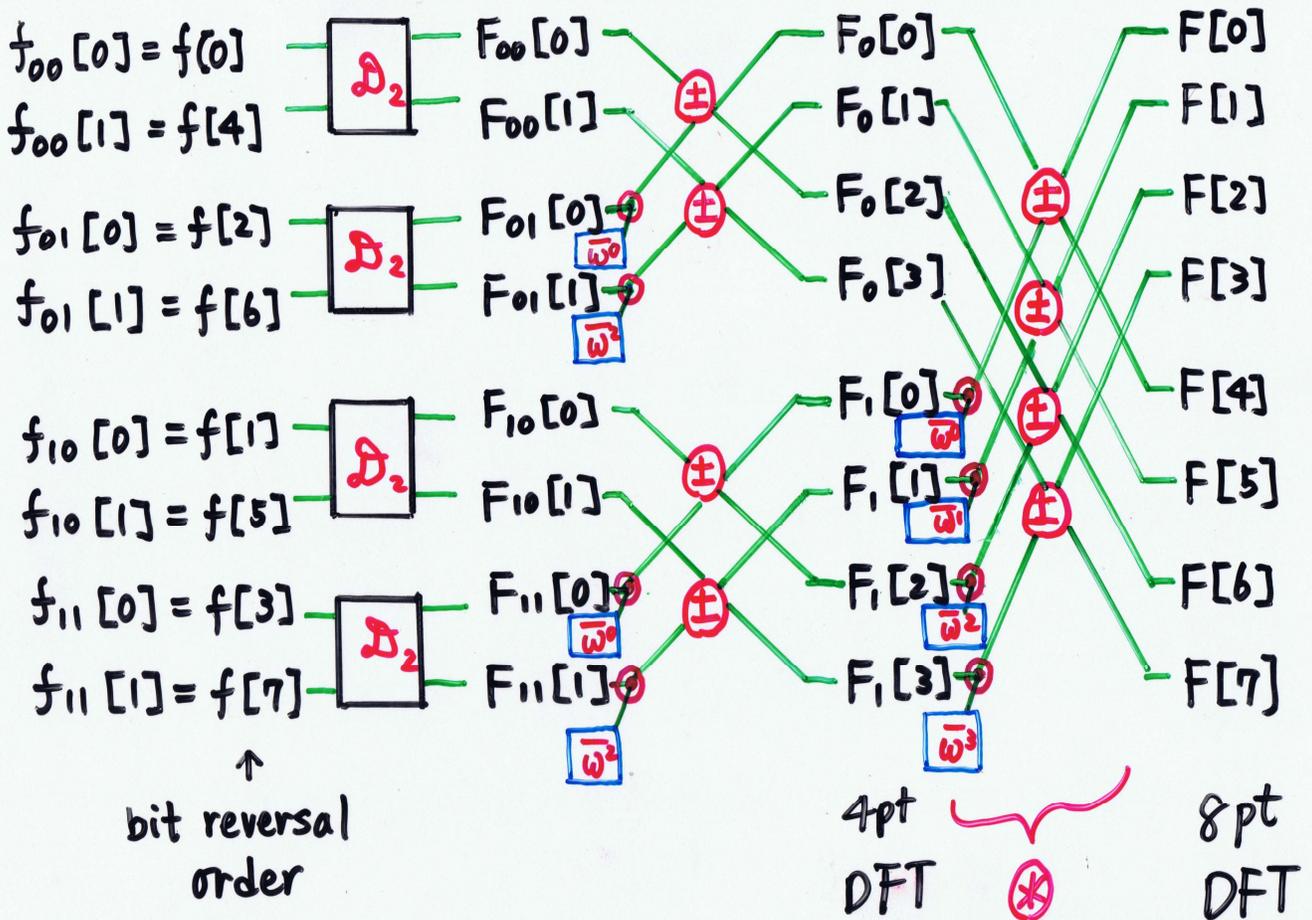
Butterfly Op.



Multiplier



Full level FFT (N=8)



$$\omega_4^1 = \omega_8^2$$

- The essence of the Covley - Tukey alg. (1965)
 - { reordering stage (even-odd, bit reversal)
 - { combine stage (butterfly op's)
- This is a bottom-up recursive procedure:
 $2 \text{ pt} \rightarrow 4 \text{ pt} \rightarrow \dots \rightarrow 2^{n-1} \text{ pt} \rightarrow 2^n \text{ pt}$

Bit Reversal Operation

• If $l = b_{m-1} b_{m-2} \dots b_0$ (binary expansion of l),
then $\bar{l} = b_0 b_1 \dots b_{m-2} b_{m-1}$ is called
the **bit-reversed** number of l .

• If $f[l] = f[b_{m-1} \dots b_0]$, then D_2 op.
at the bottom level is done between
 $f_{b_0 b_1 \dots b_{m-2}}[0]$ & $f_{b_0 b_1 \dots b_{m-2}}[1]$.

l	$b_2 b_1 b_0$	$b_0 b_1 b_2$	\bar{l}	D_2 pair
0	0 0 0	0 0 0	0	$f_{00}[0]$
1	0 0 1	1 0 0	4	$f_{10}[0]$
2	0 1 0	0 1 0	2	$f_{01}[0]$
3	0 1 1	1 1 0	6	$f_{11}[0]$
4	1 0 0	0 0 1	1	$f_{00}[1]$
5	1 0 1	1 0 1	5	$f_{10}[1]$
6	1 1 0	0 1 1	3	$f_{01}[1]$
7	1 1 1	1 1 1	7	$f_{11}[1]$

* The operation counts of the DIT alg.

- $\exists \log_2 N = n$ levels
- at each level, need $N/2$ butterfly op.
- Hence,

$$\# \text{ ops.} = \left. \begin{array}{l} \frac{N}{2} \log_2 N \text{ (}\mathbb{C}\text{-multi.'s)} \\ + N \log_2 N \text{ (}\mathbb{C}\text{-add.'s)} \end{array} \right\} \approx O(N \log_2 N)$$

★ FFT via Matrix Factorization

The whole thing can be viewed as a clever **matrix factorization!**

$$(*) \quad \mathbb{F} = W_N^* \mathbb{f} = \begin{bmatrix} I_{N/2} & \Omega_{N/2} \\ I_{N/2} & -\Omega_{N/2} \end{bmatrix} \begin{bmatrix} \mathbb{f}_0 \\ \mathbb{f}_1 \end{bmatrix}$$

where

→ represents the butterfly op's.

$$\Omega_m := \text{diag}(\omega_{2m}^0, \omega_{2m}^{-1}, \dots, \omega_{2m}^{-(m-1)}) \in \mathbb{C}^{m \times m}$$

$$(**) \quad \begin{cases} \mathbb{F}_0 = W_{N/2}^* \mathbb{f}_0 = \begin{bmatrix} I_{N/4} & \Omega_{N/4} \\ I_{N/4} & -\Omega_{N/4} \end{bmatrix} \begin{bmatrix} \mathbb{f}_{00} \\ \mathbb{f}_{01} \end{bmatrix} \\ \mathbb{F}_1 = W_{N/2}^* \mathbb{f}_1 = \begin{bmatrix} I_{N/4} & \Omega_{N/4} \\ I_{N/4} & -\Omega_{N/4} \end{bmatrix} \begin{bmatrix} \mathbb{f}_{10} \\ \mathbb{f}_{11} \end{bmatrix} \end{cases}$$

Combining (*) & (**), we have:

$$\mathbb{F} = \begin{bmatrix} I_{N/2} & \Omega_{N/2} \\ I_{N/2} & -\Omega_{N/2} \end{bmatrix} \begin{bmatrix} I_{N/4} & \Omega_{N/4} & 0 \\ I_{N/4} & -\Omega_{N/4} & 0 \\ 0 & 0 & I_{N/4} & \Omega_{N/4} \\ 0 & 0 & I_{N/4} & -\Omega_{N/4} \end{bmatrix} \begin{bmatrix} \mathbb{f}_{00} \\ \mathbb{f}_{01} \\ \mathbb{f}_{10} \\ \mathbb{f}_{11} \end{bmatrix}$$

This can be further repeated recursively.

Let $B_{2m} := \begin{bmatrix} I_m & \Omega_m \\ I_m & -\Omega_m \end{bmatrix}$, then → $\begin{bmatrix} 1 & 1 \\ 1 & -1 \end{bmatrix}_{2 \times 2}$

$$\mathbb{F} = B_N \begin{bmatrix} B_{N/2} & 0 \\ 0 & B_{N/2} \end{bmatrix} \begin{bmatrix} B_{N/4} & 0 & 0 & 0 \\ 0 & B_{N/4} & 0 & 0 \\ 0 & 0 & B_{N/4} & 0 \\ 0 & 0 & 0 & B_{N/4} \end{bmatrix} \dots \begin{bmatrix} B_2 & 0 & \dots & 0 \\ 0 & B_2 & & \vdots \\ \vdots & & \ddots & 0 \\ 0 & \dots & 0 & B_2 \end{bmatrix} \mathbb{P}^T \mathbb{f}$$

↑ bit-reversal op.

almost diagonal!

• How about D_N^{-1} ? (inverse FFT)

\Rightarrow Can use the same routine of the forward FFT.

Why? $f[l] = \frac{1}{N} \sum_{k=0}^{N-1} F[k] \omega_N^{kl}$ ← Note $\frac{1}{N}$ factor coming from the def. of D_N for this lecture.

$\Leftrightarrow N \overline{f[l]} = \sum_{k=0}^{N-1} \overline{F[k]} \omega_N^{-kl}$

$\Leftrightarrow f[l] = \frac{1}{N} \left[\sum_{k=0}^{N-1} \overline{F[k]} \omega_N^{-kl} \right]$

$\Leftrightarrow f = \frac{1}{N} \overline{\text{FFT}(\overline{F})}$ ///

★ FFT of a single \mathbb{R} -valued vector

The FFT alg. we have discussed so far is for a 1D array of complex numbers.

Can we speed up the computation of DFT for a single \mathbb{R} -valued vector?

\Rightarrow Yes! But how?

\Rightarrow Split it into even & odd indices as

$g[l] = f[2l] + i f[2l+1], \quad l=0, 1, \dots, \frac{N}{2}-1.$

↓ FFT of length $N/2$

$G[k] = \underbrace{F_0[k]}_{\in \mathbb{C}} + i \underbrace{F_1[k]}_{\in \mathbb{C}}, \quad k=0, 1, \dots, \frac{N}{2}-1.$

$F_0[k] = \sum_{l=0}^{N/2-1} f[2l] \omega_{N/2}^{-kl}, \quad F_1[k] = \sum_{l=0}^{N/2-1} f[2l+1] \omega_{N/2}^{-kl}$

Can we reconstruct $F \in \mathbb{C}^N$ from $G \in \mathbb{C}^{N/2}$?

Yes! To do so, check:

$$\begin{aligned} D_N\{f\}[k] &= F[k] = \sum_{l=0}^{N-1} f[l] \omega_N^{-kl} \\ &= \sum_{l=0}^{N/2-1} f[2l] \omega_{N/2}^{-kl} + \omega_N^{-k} \sum_{l=0}^{N/2-1} f[2l+1] \omega_{N/2}^{-kl} \end{aligned}$$

$$= F_0[k] + \omega_N^{-k} F_1[k], \quad k=0, 1, \dots, N-1$$

But, F_0 & F_1 are $N/2$ -periodic!

$$\text{So, } G[\frac{N}{2}-k] = \overline{F_0[k] + i F_1[k]}$$

$$\text{since } F_j[\frac{N}{2}-k] = F_j[-k] = \overline{F_j[k]}, \quad j=0, 1.$$

$$\begin{aligned} \text{So, } F[k] &= F_0[k] + \omega_N^{-k} F_1[k] \\ &= \frac{1}{2} \{ G[k] + \overline{G[\frac{N}{2}-k]} \} - \frac{i}{2} \{ G[k] - \overline{G[\frac{N}{2}-k]} \} \omega_N^{-k} \quad (*) \end{aligned}$$

Note $\{G[k]\}_{k=0}^{N/2-1}$ is $N/2$ -periodic.

Remark

If $f \in \mathbb{R}^N$, then

$F[0]$ = the DC comp. $\in \mathbb{R}$

$F[\frac{N}{2}]$ = the Nyquist comp. $\in \mathbb{R}$

the highest freq. comp.

So we can pack $F[0] \leftarrow F[0] + i F[\frac{N}{2}] \in \mathbb{C}$.
 $F[k] \in \mathbb{C}, k=1, \dots, N/2-1$. No need to store $F[k]$ for
 $k=N/2+1, \dots, N-1$ thanks to the symmetry.

\Rightarrow $N/2$ \mathbb{C} numbers as the output!