

Basics of Analytic Signals

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Motivation

- Many natural and man-made signals exhibit *time-varying frequencies* (e.g., chirps, FM radio waves).
- Characterization and analysis of such a signal, $u(t)$, based on *instantaneous amplitude* $a(t)$, *instantaneous phase* $\phi(t)$, and *instantaneous frequency* $\omega(t) := \phi'(t)$, are very important:

$$u(t) = a(t) \cos \phi(t).$$

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Analytic Signal

- It is convenient to use a complexified version of the signal whose real part is a given real-valued signal $u(t)$.
- Given $u(t)$, however, there are infinitely many ways to define the instantaneous amplitude and phase (IAP) pairs so that

$$u(t) = a(t) \cos \phi(t).$$

- This is due to the arbitrariness of the complexified version of u , i.e.,

$$f(t) = u(t) + i v(t)$$

where $v(t)$ is an arbitrary real-valued signal; yet this yields the IAP representation of $u(t)$ via

$$a(t) = \sqrt{u^2(t) + v^2(t)}, \quad \phi(t) = \arctan \frac{v(t)}{u(t)}.$$

- The *instantaneous frequency* is defined as

$$\omega(t) := \frac{d\phi}{dt} = \frac{u(t)v'(t) - u'(t)v(t)}{u^2(t) + v^2(t)}.$$

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- Gabor (1946) proposed to use the *the Hilbert transform* of $u(t)$ as $v(t)$, and called the complex-valued $f(t)$ an *analytic signal*.
- Vakman (1972) proved that $v(t)$ must be of the Hilbert transform of $u(t)$ if we impose some a priori physical assumptions:
 - $v(t)$ must be derived from $u(t)$.
 - Amplitude continuity: a small change in $u \implies$ a small change in $a(t)$.
 - Phase independence of scale: if $cu(t)$, $c \in \mathbb{R}$ arbitrary scalar, then the phase does not change from that of $u(t)$ and its amplitude becomes c times that of $u(t)$.
 - Harmonic correspondence: if $u(t) = a_0 \cos(\omega_0 t + \phi_0)$, then $a(t) = a_0$, $\phi(t) = \omega_0 t + \phi_0$.

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Analytic Signal ...

- For simplicity, we assume that our signals are 2π -periodic in $\theta \in [-\pi, \pi)$.
- Hence, we work on the unit circle and unit disk \mathbb{D} in $\mathbb{C} = \mathbb{R}^2$.
- Note that the signals over $\mathbb{R} = (-\infty, \infty)$ can be treated similarly by considering the real axis and the upper half plane of \mathbb{C} .
- The analytic signal of a given signal $u(\theta) \in \mathbb{R}$ is often and simply obtained via the *Hilbert transform*:

$$f(\theta) = u(\theta) + i\mathcal{H}u(\theta), \quad \mathcal{H}u(\theta) := \frac{1}{2\pi} \text{pv} \int_{-\pi}^{\pi} u(\tau) \cot \frac{\theta - \tau}{2} d\tau.$$

- Note that

$$u(\theta) = \frac{a_0}{2} + \sum_{k \geq 1} (a_k \cos k\theta + b_k \sin k\theta) \Rightarrow \mathcal{H}u(\theta) = \sum_{k \geq 1} (a_k \sin k\theta - b_k \cos k\theta).$$

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Analytic Signal ...

We can gain a deeper insight by viewing this as *the boundary value* of an *analytic function* $F(z)$ where

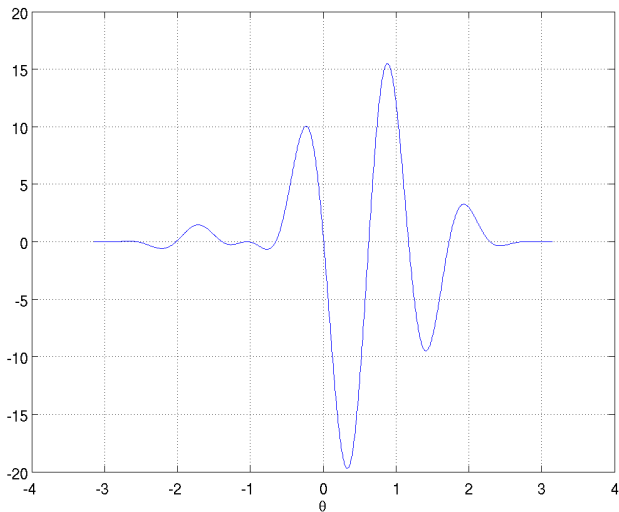
$$F(z) := U(z) + i\tilde{U}(z), \quad z \in \mathbb{D},$$

where

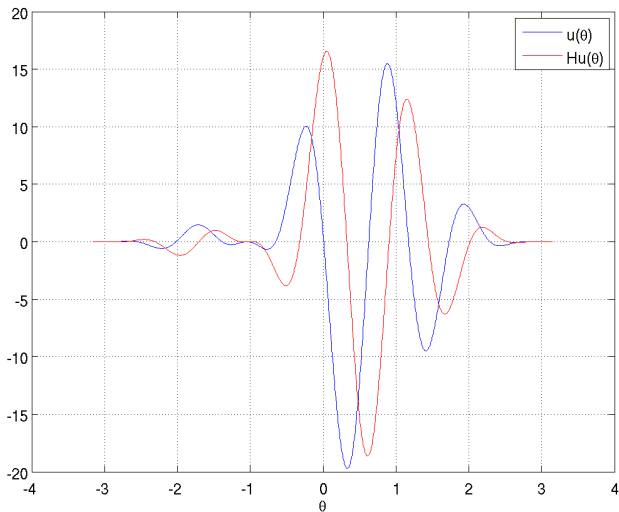
$$U(z) = U(re^{i\theta}) = P_r * u(\theta) = \frac{1}{2\pi} \int_{-\pi}^{\pi} \frac{1-r^2}{1-2r\cos(\theta-\tau)+r^2} u(\tau) d\tau,$$
$$\tilde{U}(z) = \tilde{U}(re^{i\theta}) = Q_r * u(\theta) = \frac{1}{2\pi} \int_{-\pi}^{\pi} \frac{2r\sin(\theta-\tau)}{1-2r\cos(\theta-\tau)+r^2} u(\tau) d\tau.$$

In other words, the original signal $u(\theta) = U(e^{i\theta})$ is *the boundary value of the harmonic function U on $\partial\mathbb{D}$* , which is constructed by *the Poisson integral*. \tilde{U} and $Q_r(\theta)$ are referred to as the conjugate harmonic function and the conjugate Poisson kernel, respectively.

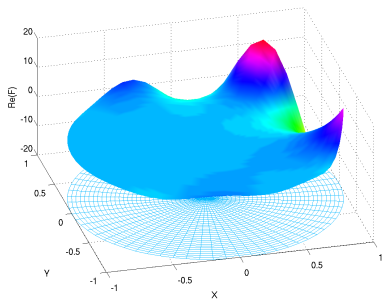
Analytic Signal ... An Example: $u(\theta)$



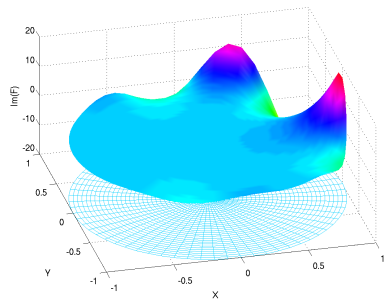
Analytic Signal ... An Example: $u(\theta)$ and $\mathcal{H}u(\theta)$



Analytic Signal ... An Example: $U(z)$ and $\tilde{U}(z)$



(a) $U(z)$

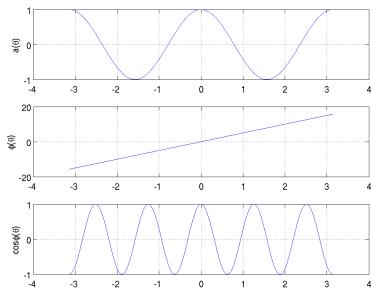


(b) $\tilde{U}(z)$

Analytic Signal ...

Even if we use the analytic signal, its IAP representation is not unique as shown by Cohen, Loughlin, and Vakman (1999):

- $f(\theta) = a(\theta)e^{i\phi(\theta)}$, where $a(\theta) = u(\theta) \cos \phi(\theta) + v(\theta) \sin \phi(\theta)$ may be negative though $\phi(\theta)$ is continuous;
- $f(\theta) = |a(\theta)|e^{i(\phi(\theta)+\pi\alpha(\theta))}$, where $\alpha(\theta)$ is an appropriate phase function, which may be discontinuous.

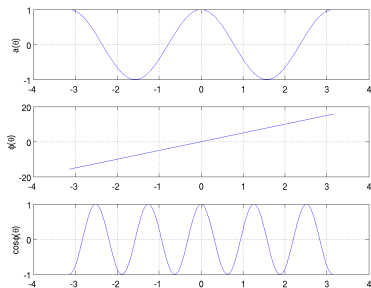


(a) Continuous phase

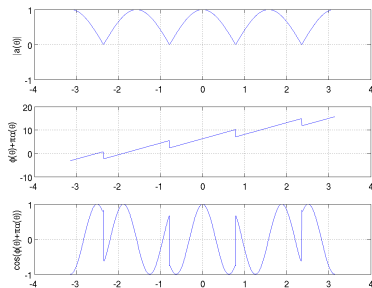
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(a) Continuous phase



(b) Nonnegative amplitude